

NPN 9 GHz wideband transistor

BFG520; BFG520/X; BFG520/XR

FEATURES

- High power gain
- Low noise figure
- High transition frequency
- Gold metallization ensures excellent reliability.

DESCRIPTION

NPN silicon planar epitaxial transistors, intended for applications in the RF frontend in the GHz range, such as analog and digital cellular telephones, cordless telephones (CT1, CT2, DECT, etc.), radar detectors, pagers and satellite TV tuners (SATV) and repeater amplifiers in fibre-optic systems.

The transistors are encapsulated in 4-pin, dual-emitter plastic SOT143 and SOT143R envelopes.

PINNING

PIN	DESCRIPTION
BFG520 (Fig.1) Code: N36	
1	collector
2	base
3	emitter
4	emitter
BFG520/X (Fig.1) Code: N42	
1	collector
2	emitter
3	base
4	emitter
BFG520/XR (Fig.2) Code: N48	
1	collector
2	emitter
3	base
4	emitter

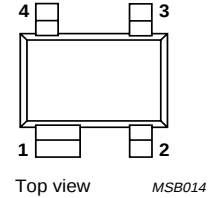


Fig.1 SOT143B.

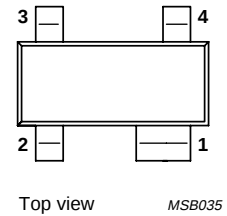


Fig.2 SOT143R.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	–	–	20	V
V_{CEO}	collector-emitter voltage	open base	–	–	15	V
I_C	DC collector current		–	–	70	mA
P_{tot}	total power dissipation	up to $T_s = 88\text{ °C}$; note 1	–	–	300	mW
h_{FE}	DC current gain	$I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $T_j = 25\text{ °C}$	60	120	250	
C_{re}	feedback capacitance	$I_C = 0$; $V_{CB} = 6\text{ V}$; $f = 1\text{ MHz}$	–	0.3	–	pF
f_T	transition frequency	$I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 1\text{ GHz}$; $T_{amb} = 25\text{ °C}$	–	9	–	GHz
G_{UM}	maximum unilateral power gain	$I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 900\text{ MHz}$; $T_{amb} = 25\text{ °C}$	–	19	–	dB
		$I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 2\text{ GHz}$; $T_{amb} = 25\text{ °C}$	–	13	–	dB
$ S_{21} ^2$	insertion power gain	$I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 900\text{ MHz}$; $T_{amb} = 25\text{ °C}$	17	18	–	dB
F	noise figure	$\Gamma_s = \Gamma_{opt}$; $I_C = 5\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 900\text{ MHz}$; $T_{amb} = 25\text{ °C}$	–	1.1	1.6	dB
		$\Gamma_s = \Gamma_{opt}$; $I_C = 20\text{ mA}$; $V_{CE} = 6\text{ V}$; $f = 900\text{ MHz}$; $T_{amb} = 25\text{ °C}$	–	1.6	2.1	dB
		$\Gamma_s = \Gamma_{opt}$; $I_C = 5\text{ mA}$; $V_{CE} = 8\text{ V}$; $f = 2\text{ GHz}$; $T_{amb} = 25\text{ °C}$	–	1.9	–	dB

LIMITING VALUES

In accordance with the Absolute Maximum System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	–	20	V
V_{CEO}	collector-emitter voltage	open base	–	15	V
V_{EBO}	emitter-base voltage	open collector	–	2.5	V
I_C	DC collector current		–	70	mA
P_{tot}	total power dissipation	up to $T_s = 88\text{ °C}$; note 1	–	300	mW
T_{stg}	storage temperature		–65	150	°C
T_j	junction temperature		–	175	°C

THERMAL RESISTANCE

SYMBOL	PARAMETER	CONDITIONS	THERMAL RESISTANCE
$R_{th\ j-s}$	thermal resistance from junction to soldering point	up to $T_s = 88\text{ °C}$; note 1	290 K/W

CHARACTERISTICST_j = 25 °C unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I _{CBO}	collector cut-off current	I _E = 0; V _{CB} = 6 V	–	–	50	nA
h _{FE}	DC current gain	I _C = 20 mA; V _{CE} = 6 V	60	120	250	
C _e	emitter capacitance	I _C = i _c = 0; V _{EB} = 0.5 V; f = 1 MHz	–	1	–	pF
C _c	collector capacitance	I _E = i _e = 0; V _{CB} = 6 V; f = 1 MHz	–	0.6	–	pF
C _{re}	feedback capacitance	I _C = 0; V _{CB} = 6 V; f = 1 MHz	–	0.3	–	pF
f _T	transition frequency	I _C = 20 mA; V _{CE} = 6 V; f = 1 GHz; T _{amb} = 25 °C	–	9	–	GHz
G _{UM}	maximum unilateral power gain (note 1)	I _C = 20 mA; V _{CE} = 6 V; f = 900 MHz; T _{amb} = 25 °C	–	19	–	dB
		I _C = 20 mA; V _{CE} = 6 V; f = 2 GHz; T _{amb} = 25 °C	–	13	–	dB
S ₂₁ ²	insertion power gain	I _C = 20 mA; V _{CE} = 6 V; f = 900 MHz; T _{amb} = 25 °C	17	18	–	dB
F	noise figure	Γ _S = Γ _{opt} ; I _C = 5 mA; V _{CE} = 6 V; f = 900 MHz; T _{amb} = 25 °C	–	1.1	1.6	dB
		Γ _S = Γ _{opt} ; I _C = 20 mA; V _{CE} = 6 V; f = 900 MHz; T _{amb} = 25 °C	–	1.6	2.1	dB
		Γ _S = Γ _{opt} ; I _C = 5 mA; V _{CE} = 6 V; f = 2 GHz; T _{amb} = 25 °C	–	1.9	–	dB
P _{L1}	output power at 1 dB gain compression	I _C = 20 mA; V _{CE} = 6 V; R _L = 50 Ω; f = 900 MHz; T _{amb} = 25 °C	–	17	–	dBm
ITO	third order intercept point	note 2	–	26	–	dBm
V _o	output voltage	note 3	–	275	–	mV
d ₂	second order intermodulation distortion	I _C = 20 mA; V _{CE} = 6 V; V _o = 75 mV; T _{amb} = 25 °C; f _(p+q) = 810 MHz	–	–50	–	dB

Notes

1. G_{UM} is the maximum unilateral power gain, assuming S₁₂ is zero and

$$G_{UM} = 10 \log \left(\frac{|S_{21}|^2}{(1 - |S_{11}|^2)(1 - |S_{22}|^2)} \right) \text{dB.}$$

2. I_C = 20 mA; V_{CE} = 6 V; R_L = 50 Ω; f = 900 MHz; T_{amb} = 25 °C;
f_p = 900 MHz; f_q = 902 MHz;
measured at f_(2p-q) = 898 MHz and f_(2q-p) = 904 MHz.
3. d_{im} = –60 dB (DIN 45004B);
V_p = V_o; V_q = V_o – 6 dB; V_r = V_o – 6 dB;
f_p = 795.25 MHz; f_q = 803.25 MHz; f_r = 805.25 MHz;
measured at f_(p+q-r) = 793.25 MHz

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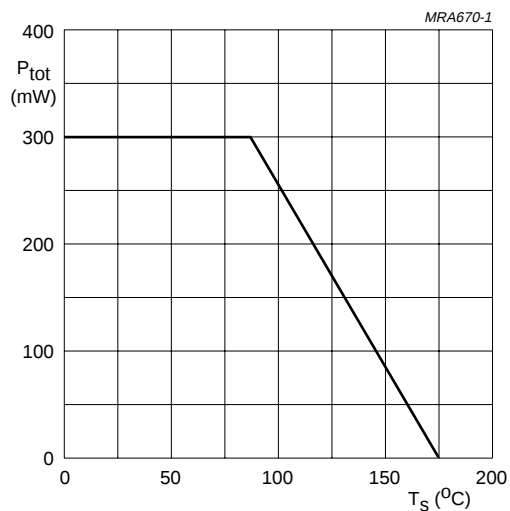
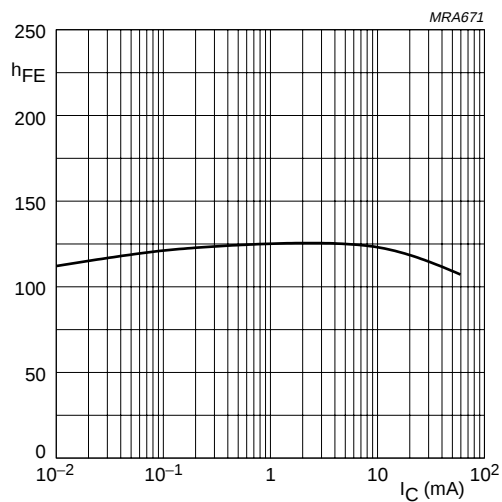
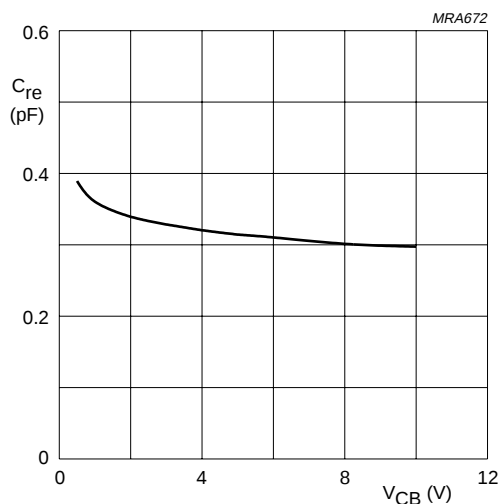


Fig.3 Power derating curve.



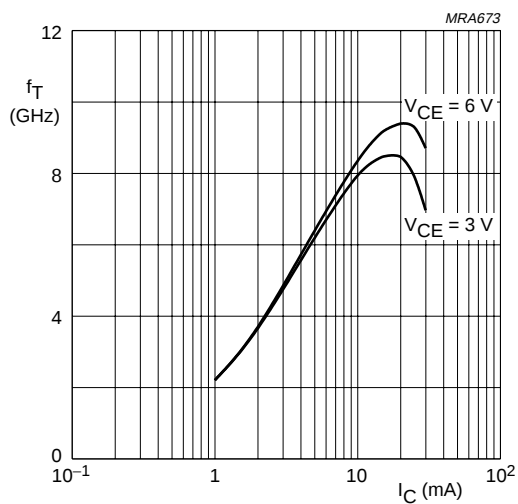
$V_{CE} = 6 \text{ V}$; $T_J = 25 \text{ }^{\circ}\text{C}$.

Fig.4 DC current gain as a function of collector current.



$I_C = 0$; $f = 1 \text{ MHz}$.

Fig.5 Feedback capacitance as a function of collector-base voltage.



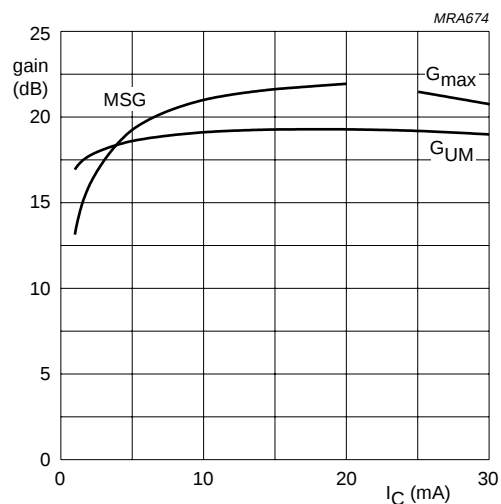
$f = 1 \text{ GHz}$; $T_{amb} = 25 \text{ }^{\circ}\text{C}$.

Fig.6 Transition frequency as a function of collector current.

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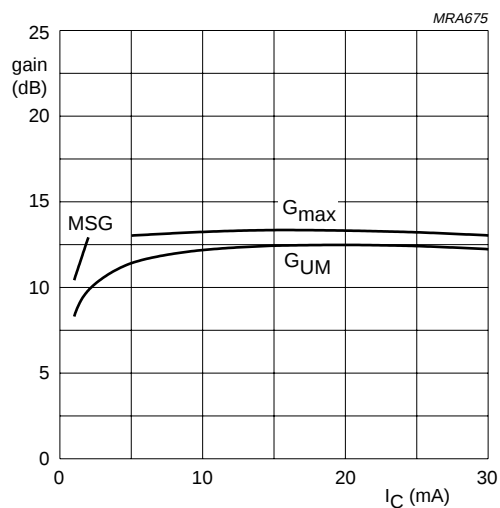
BFG520; BFG520/X; BFG520/XR

In Figs 7 to 10, G_{UM} = maximum unilateral power gain; MSG = maximum stable gain; G_{max} = maximum available gain.



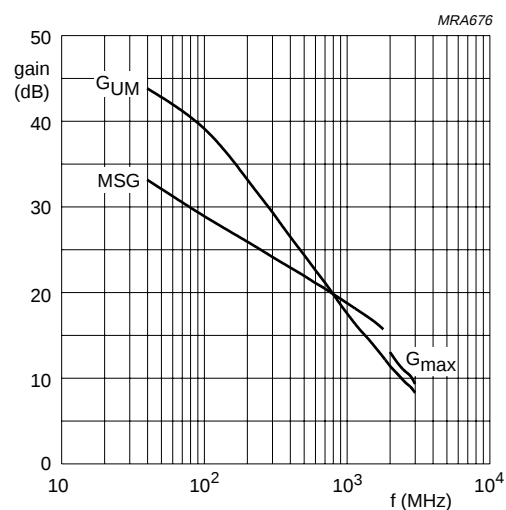
$V_{CE} = 6 \text{ V}$; $f = 900 \text{ MHz}$; $T_{amb} = 25 \text{ °C}$.

Fig.7 Gain as a function of collector current.



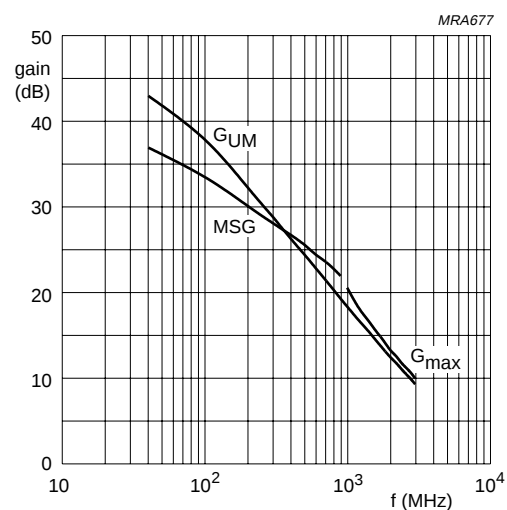
$V_{CE} = 6 \text{ V}$; $f = 2 \text{ GHz}$; $T_{amb} = 25 \text{ °C}$.

Fig.8 Gain as a function of collector current.



$I_C = 5 \text{ mA}$; $V_{CE} = 6 \text{ V}$; $T_{amb} = 25 \text{ °C}$.

Fig.9 Gain as a function of frequency.



$I_C = 20 \text{ mA}$; $V_{CE} = 6 \text{ V}$; $T_{amb} = 25 \text{ °C}$.

Fig.10 Gain as a function of frequency.

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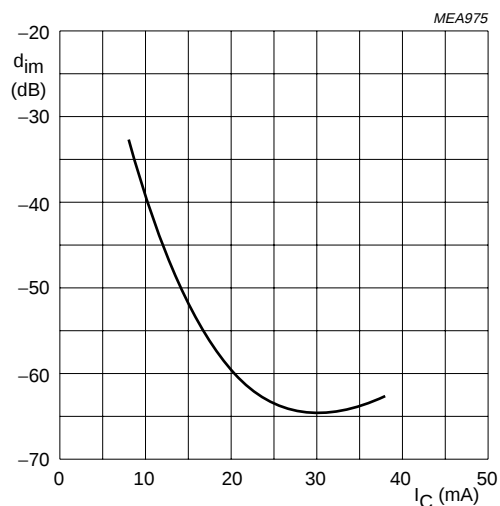


Fig.11 Intermodulation distortion as a function of collector current.

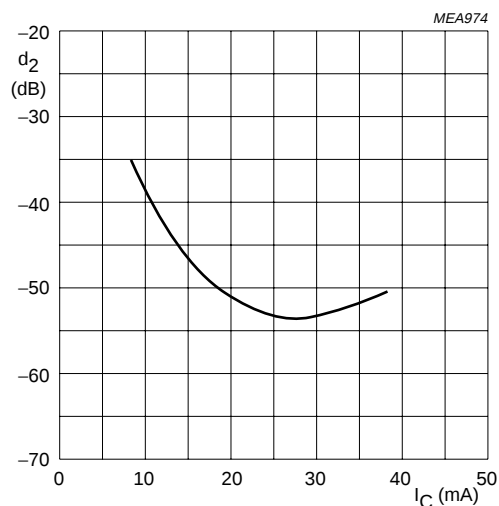
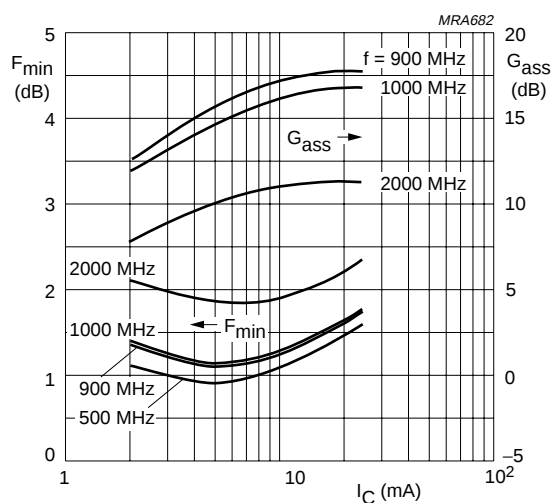
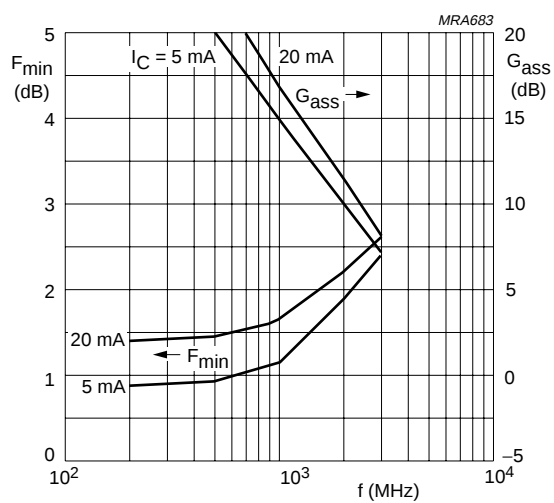


Fig.12 Second order intermodulation distortion as a function of collector current.



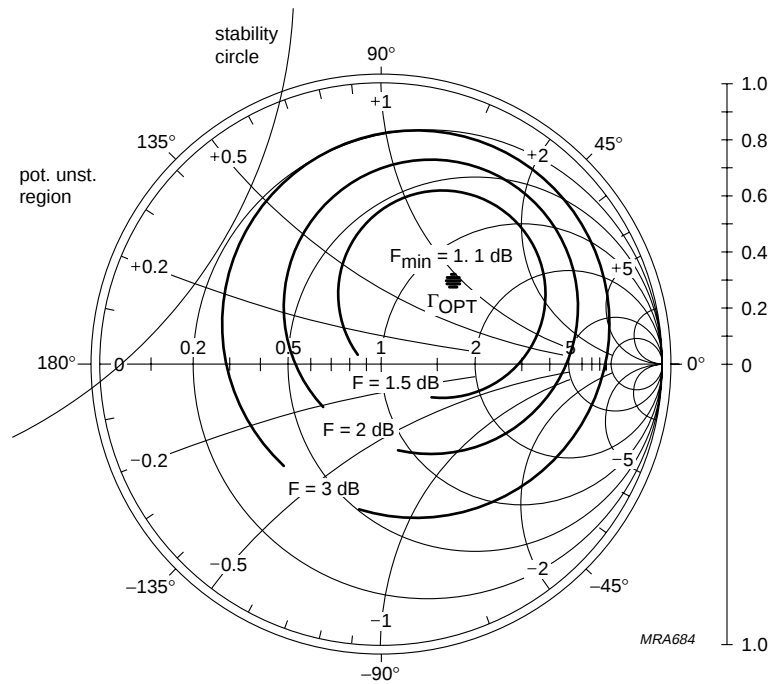
$V_{CE} = 6 \text{ V}$; $T_{amb} = 25 \text{ }^{\circ}\text{C}$.

Fig.13 Minimum noise figure and associated available gain as functions of collector current.



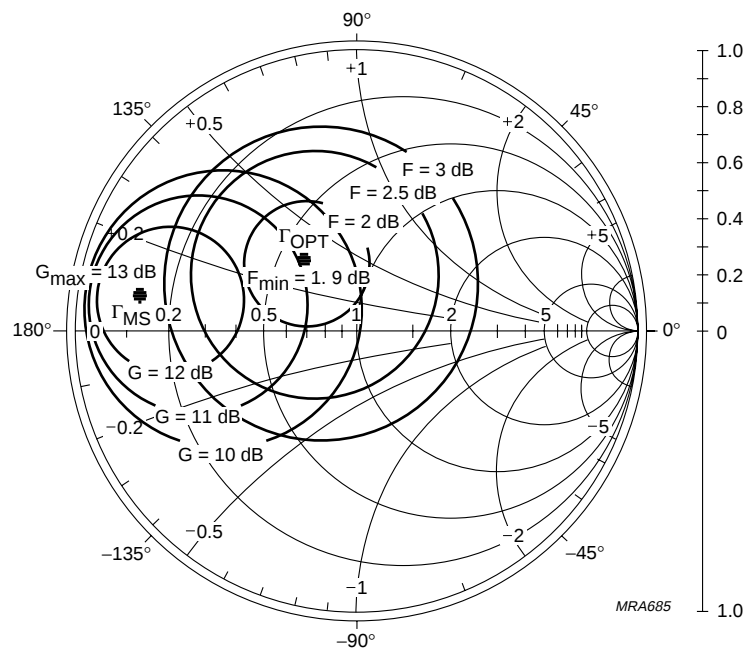
$V_{CE} = 6 \text{ V}$; $T_{amb} = 25 \text{ }^{\circ}\text{C}$.

Fig.14 Minimum noise figure and associated available gain as functions of frequency.



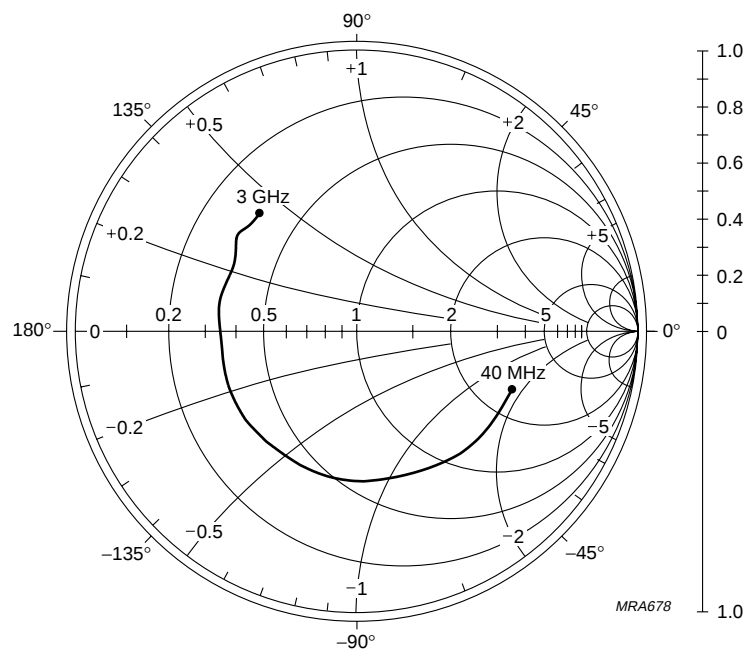
$I_C = 5 \text{ mA}$; $V_{CE} = 6 \text{ V}$;
 $f = 900 \text{ MHz}$; $Z_0 = 50 \Omega$.

Fig.15 Noise circle figure.



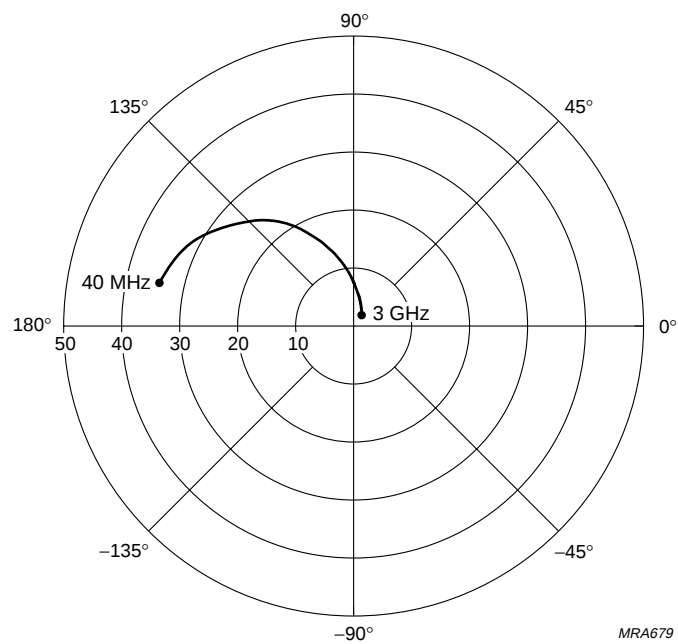
$I_C = 5 \text{ mA}$; $V_{CE} = 6 \text{ V}$;
 $f = 2 \text{ GHz}$; $Z_0 = 50 \Omega$.

Fig.16 Noise circle figure.



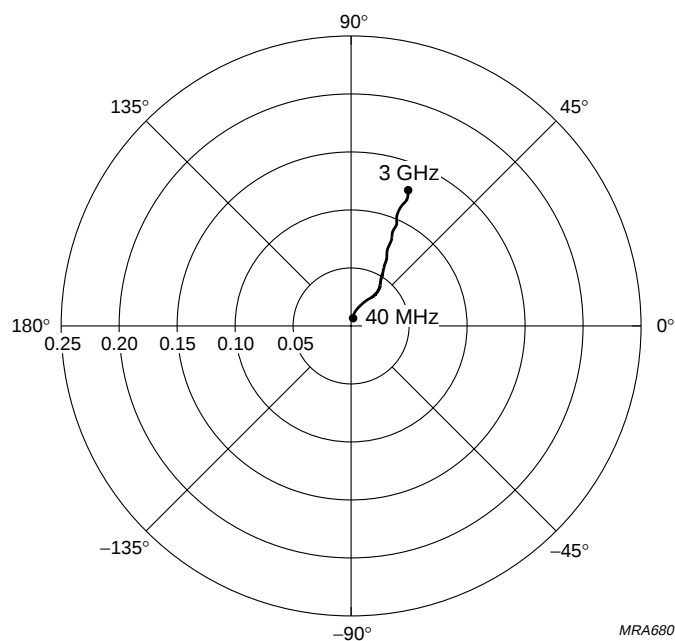
$I_C = 20 \text{ mA}$; $V_{CE} = 6 \text{ V}$.
 $Z_0 = 50 \Omega$.

Fig.17 Common emitter input reflection coefficient (S_{11}).



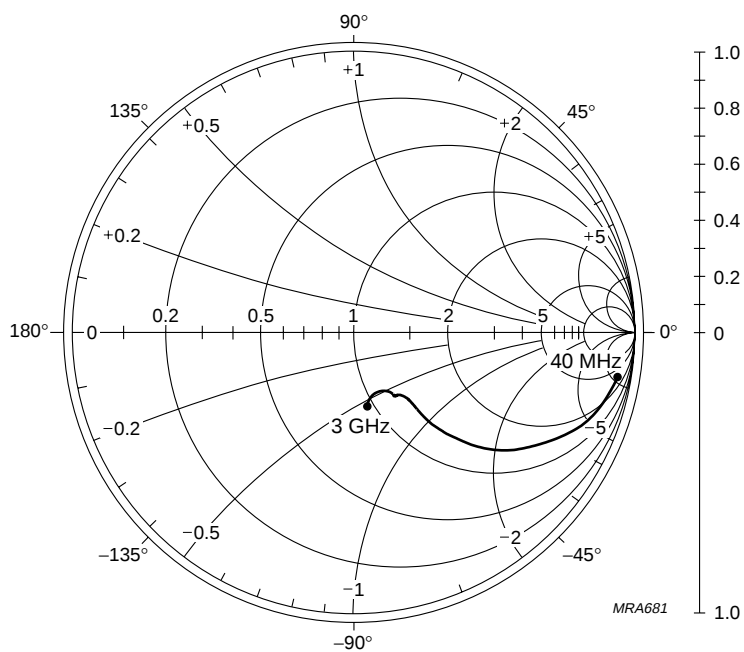
$I_C = 20 \text{ mA}$; $V_{CE} = 6 \text{ V}$.

Fig.18 Common emitter forward transmission coefficient (S_{21}).



$I_C = 20 \text{ mA}$; $V_{CE} = 6 \text{ V}$.

Fig.19 Common emitter reverse transmission coefficient (S_{12}).



$I_C = 20 \text{ mA}$; $V_{CE} = 6 \text{ V}$.
 $Z_0 = 50 \Omega$.

Fig.20 Common emitter output reflection coefficient (S_{22}).

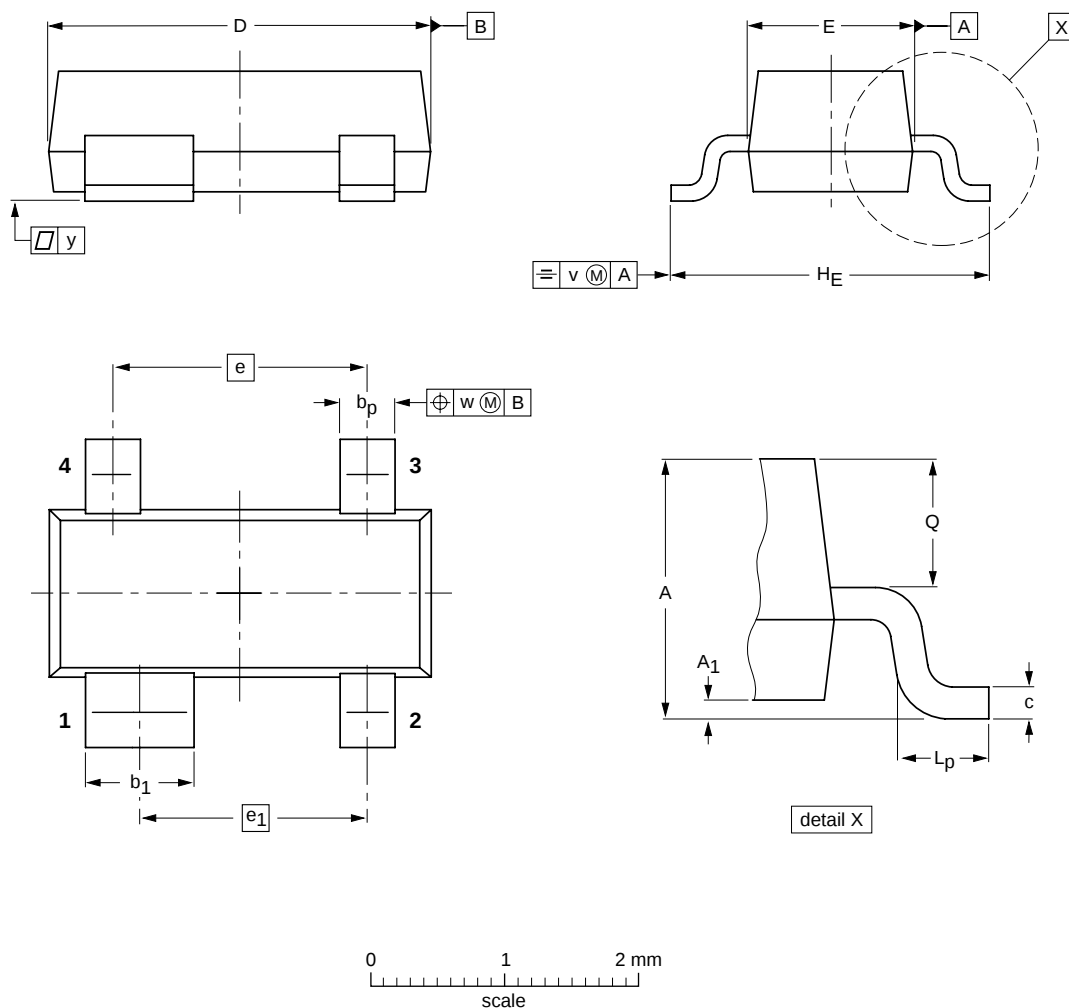
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PACKAGE OUTLINES

Plastic surface mounted package; 4 leads

SOT143B



DIMENSIONS (mm are the original dimensions)

UNIT	A	A ₁ max	b _p	b ₁	c	D	E	e	e ₁	H _E	L _p	Q	v	w	y
mm	1.1 0.9	0.1	0.48 0.38	0.88 0.78	0.15 0.09	3.0 2.8	1.4 1.2	1.9	1.7	2.5 2.1	0.45 0.15	0.55 0.45	0.2	0.1	0.1

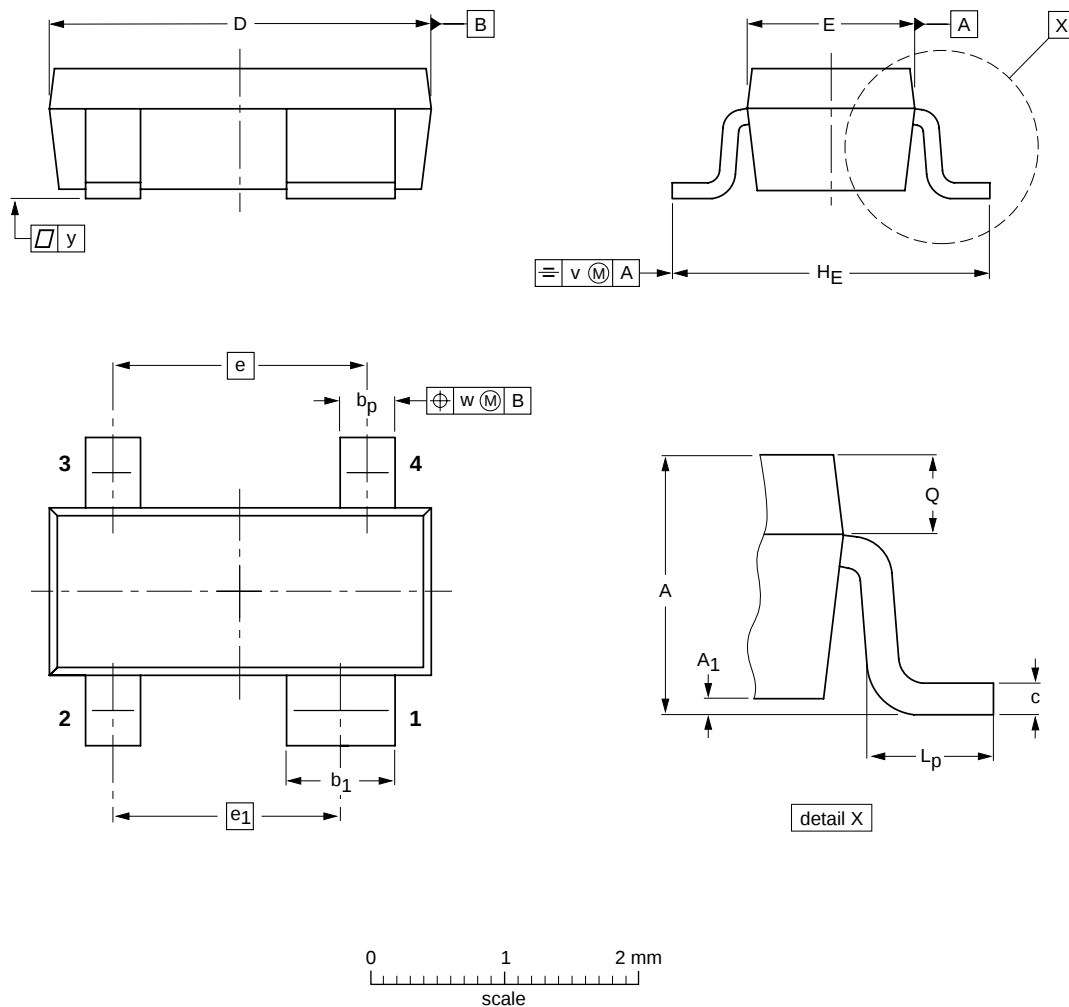
OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT143B						97-02-28

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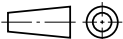
Plastic surface mounted package; reverse pinning; 4 leads

SOT143R



DIMENSIONS (mm are the original dimensions)

UNIT	A	A ₁ max	b _p	b ₁	c	D	E	e	e ₁	H _E	L _p	Q	v	w	y
mm	1.1 0.9	0.1	0.48 0.38	0.88 0.78	0.15 0.09	3.0 2.8	1.4 1.2	1.9	1.7	2.5 2.1	0.55 0.25	0.45 0.25	0.2	0.1	0.1

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT143R						97-03-10